

Detecting targets with quantum microwave signals

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Sh. Barzanjeh et al., Sci. Adv. 6, eabb0451 (2020).

Outline

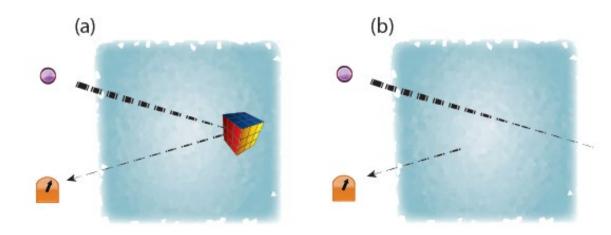
- Quantum illumination: protocol for target detection; an example of quantum sensing and its advantages
- It is a binary detection, rather than positioning, protocol; target presence or absence
- quantum advantage over classical strategies via optimal quantum state discrimination; determination of the optimal input state and optimal detection scheme
- Experimental demonstration with a microwave entangled source and digital postprocessing reproducing a phase-conjugate receiver, Sh. Barzanjeh et al., Sci. Adv. 6, eabb0451 (2020). An important step toward a <u>"quantum radar"</u>

Standard classical target detection

- single probe beam (e.g. coherent state) sent into a noisy region to detect the eventual presence of an object.
- (a) Target present ⇒ small chance a reflected signal is detected; (b) Target absent ⇒ the probe is lost and receiver sees only noise.

Typical scenario:

- i) Low reflectivity \Leftrightarrow high loss $\eta << 1$
- ii) Weak transmission \Leftrightarrow low signal $n_s <<1$
- iii) Bright background large noise $n_b >> 1$



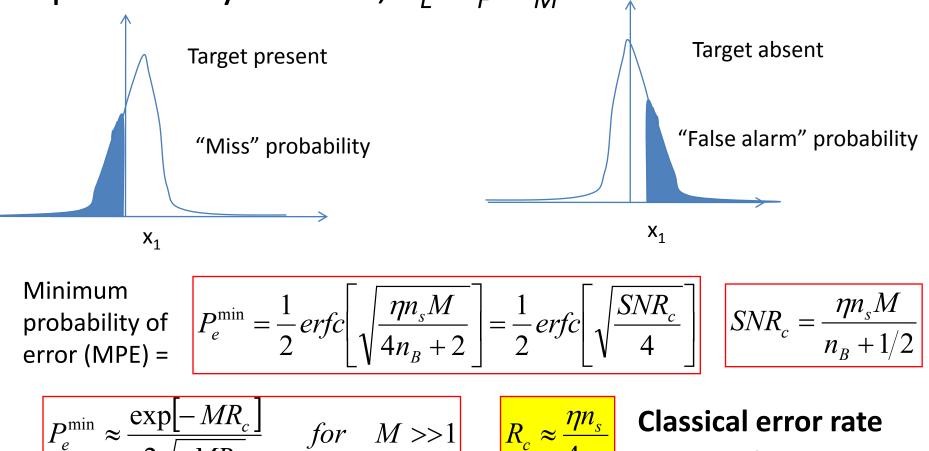
LOW SNR, i.e., typical radar detection regime

Classical binary detection problem

- Discrimination between two hypotheses H₀ (no target) and H₁ (target present)
- $H_0 \Rightarrow$ return mode a_R = thermal state with n_b mean photons and $< a_R > = 0$
- $H_1 \Rightarrow$ return mode a_R = thermal state with n_b mean photons and $\langle a_R \rangle = \sqrt{\eta} n_s$
- Optimal classical strategy: i) optimal input state = coherent state; ii) optimal detection = quadrature X_i measurement with homodyne
- send M >> 1 modes and measure $X = \sum_{i=1}^{M} X_i \implies \text{distinguishing}$ between two Gaussians shifted by $x_1^{i=1} M(\eta n_s)^{1/2}/2$

Classical binary detection problem II

• The standard strategy is to minimize the probability of error, $P_F = P_F + P_M$

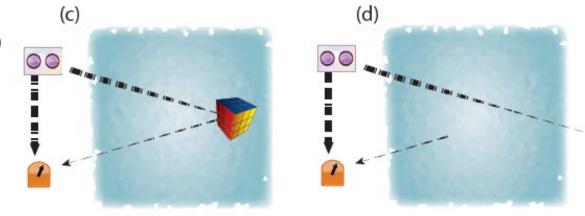


exponent

Quantum illumination protocol

- Different input state: two **maximally-entangled beams**, one is kept (idler) and the other one sent for target detection (signal).
- The reflected signal and idler are finally detected by an **appropriate joint measurement**.
- the use of an entangled source yields better performance, even though entanglement fails to survive the return trip.

S. Lloyd, Science, 321, 1043 (2008)
S.-H. Tan et al., Phys. Rev. Lett.
101, 253601 (2008).
S. Guha and B. I. Erkmen, Phys.
Rev. A 80, 052310 (2009).
C. Weedbrook et al., New J. Phys.
18 043027 (2016)



Quantum ideal case

 Optimal input state = Two-mode squeezed state of a signal and idler (G. De Palma & J. Borregaard, Phys. Rev. A 98, 012101 (2018))

$$\left|\psi\right\rangle_{SI} = \sum_{n=1}^{\infty} \sqrt{\frac{\overline{n_s}^n}{(\overline{n_s}+1)^{n+1}}} \left|n\right\rangle_S \left|n\right\rangle_I$$

- $H_0 \Rightarrow$ return mode $a_R = a_{B_i}$ in a thermal state with $n_b >>1$ mean photons
- $H_1 \Rightarrow$ return mode $a_R = \sqrt{\eta}a_s + \sqrt{1-\eta}a_B$
- Optimal binary detection (Helstrom) with multicopies (M>>1): maximum "distance" between the quantum (mixed) states related to the two hypotheses ⇒

$$P_{e}^{\min} = \frac{1}{2} \left[1 - \frac{1}{2} Tr \left| \left(\hat{\rho}_{RI}^{0} \right)^{\otimes M} - \left(\hat{\rho}_{RI}^{1} \right)^{\otimes M} \right| \right]$$

 ρ_{RI}^{j} j= 0,1 joint states of the return-idler mode system for the two hyp.

$QI \Rightarrow 6 dB gain in error exponent$

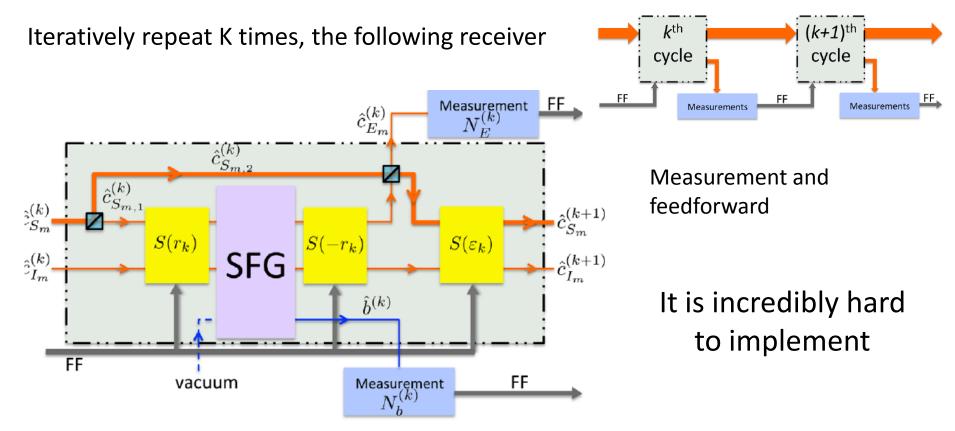
$$P_{e}^{\min} \leq \frac{1}{2} \inf_{0 \leq s \leq 1} Tr\left[\left(\hat{\rho}_{RI}^{0}\right)^{s} \left(\hat{\rho}_{RI}^{0}\right)^{1-s}\right]$$
Quantum Chernoff bound
exponentially tight at large M
$$P_{e}^{\min} \approx \frac{\exp\left[-MR_{Q}\right]}{2\sqrt{\pi MR_{Q}}} \quad for \quad M >>1$$

$$R_{Q} \approx \frac{\eta n_{s}}{n_{b}} = 4R_{C} \quad 6 \text{ dB gain}$$
Optimal quantum
error rate exponent

This is theory: How to realize such **optimal quantum receiver** ? Which experimentally feasible **detection scheme** achieves (or at least approaches) the optimal quantum error rate exponent R_o ?

Optimal proof-of-principle detection scheme (Q. Zhuang, Z. Zhang, and J. H. Shapiro, PRL 118, 040801 (2017))

The FF-SFG (feedforward sum frequency generation) scheme achieving Helstrom's optimal binary detection



SFG

$$\hat{H}_I = \hbar g \sum_{m=1}^M (\hat{b}^\dagger \hat{a}_{S_m} \hat{a}_{I_m} + \hat{b} \hat{a}_{S_m}^\dagger \hat{a}_{I_m}^\dagger),$$

Sum-frequency generation: "time-inverted" parametric down conversion: M+1 modes coherent interaction, very hard to realize

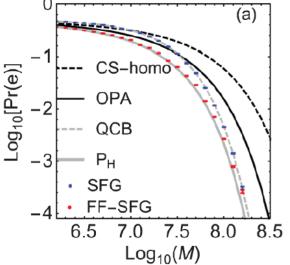
Optimal mixed-state discrimination: it is known that it is achievable only with COLLECTIVE measurements and not with LOCC strategies

The SFG, in the low-excitation limit of very small signal and idler photon number, converts the signal-idler correlation into a nonzero amplitude of the sum-frequency (pump) beam b(t)

$$C(t) \equiv \langle \hat{a}_{S_m} \hat{a}_{I_m} \rangle_t \quad \begin{aligned} C(t) &\equiv C(0) \cos(\sqrt{M}gt), \\ b(t) &= -i\sqrt{M}C(0) \sin(\sqrt{M}gt) \end{aligned}$$

SFG maps the problem to the optimal discrimination of two coherent states (Dolinar receiver)

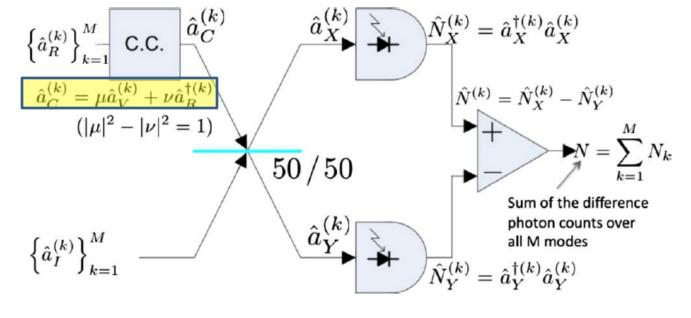
The three squeezers, the beam splitters, the measurements, and feedforward for K cycles are needed to reach this optimal low photon number regime



Phase conjugate receiver (PCR)

An easier to implement quantum receiver

S. Guha and B. I. Erkmen, Phys. Rev. A 80, 052310 (2009).



phase conjugation followed by balanced dual **photodetection**

 $H_0 \Longrightarrow < N_0 > = 0;$ $H_1 \Longrightarrow < N_1 > = 2M < a_c^+ a_1 > \neq 0$ Phase-insensitive correlations, related to signal-idler quantum correlations

$$P_{\mathrm{QI}}^{(M)} = \frac{P_F + P_M}{2} = \frac{\operatorname{erfc}\left(\sqrt{\operatorname{SNR}_{\mathrm{QI}}^{(M)}/8}\right)}{2}$$

Error probability in the *M>>1* Central Limit Theorem Gaussian limit

$$P_e^{\min} \approx \frac{\exp[-MR_{PCR}]}{2\sqrt{\pi MR_{PCR}}} \quad for \quad M >>1$$

$$R_{PCR} \approx \frac{\eta n_s}{2n_b} = 2R_C$$

3 dB gain wrt classical case

In the usual scenario:

$$) \quad \eta << 1$$

ii) low signal $n_s \ll 1$

iii) $n_b >> 1$

Not the optimal quantum 6 dB gain, but already significantly better than any classical target detection. This is the detection strategy we have implemented experimentally by digital postprocessing

Experimental demonstrations

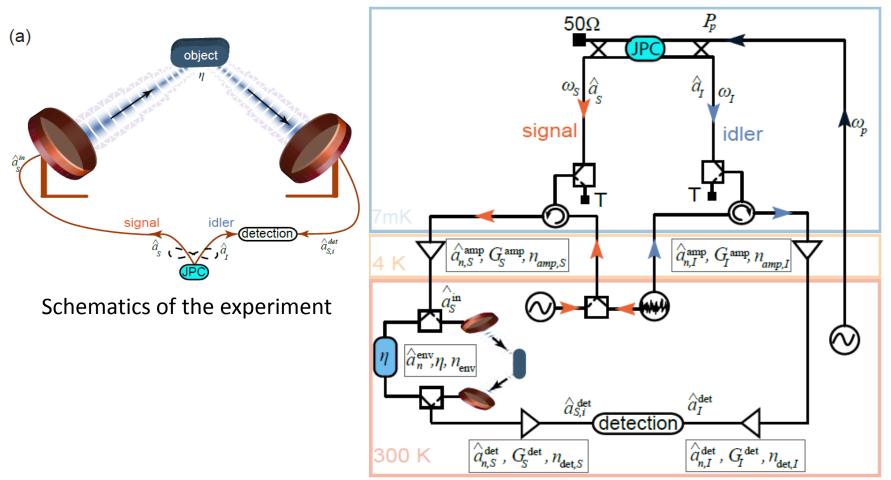
- QI has been first demonstrated at optical wavelengths, where noise has been artificially added
- [E. D. Lopaeva et al., Phys. Rev. Lett. 110, 153603 (2013); Z. Zhang et al., Phys. Rev. Lett. 114, 110506 (2015)]
- QI could be useful in radar applications at μ-waves, where one is easily in the low SNR regime
- First use of quantum entangled sources (with a Josephson parametric converter (JPC)) for target detection at μ -waves

in C. W. Sandbo Chang et al., Appl. Phys. Lett. 114, 112601 (2019), and D. Luong et al., Trans. Aerosp. Electron. Syst. 10.1109/TAES.2019.2951213 (2019).

Heterodyne detection which is not optimal, and no quantum advantage clearly demonstrated

Our demonstration of microwave QI: employing a digital version of the phase-conjugated receiver

- We again use the JPC entangled source of two-mode squeezed signal-idler beams. Generated at 7 mK
- By postprocessing heterodyne data, we digitally simulate the phase-conjugate receiver (PCR) giving a 3 dB rateexponent gain, Detection of a room-T target (up to 1 meter)
- We compare with the optimal classical detection (under the same conditions): coherent state & homodyne detection (passing through the same amplification/detection channel)

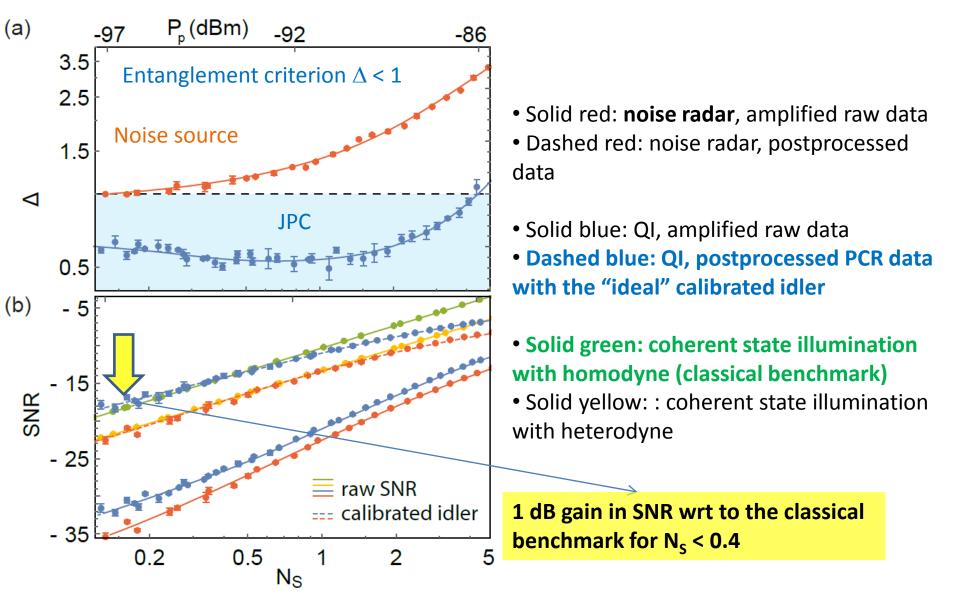


• The JPC output (or the reflected classical signals) are amplified, down-converted, heterodyned, and digitized simultaneously and independently for both channels.

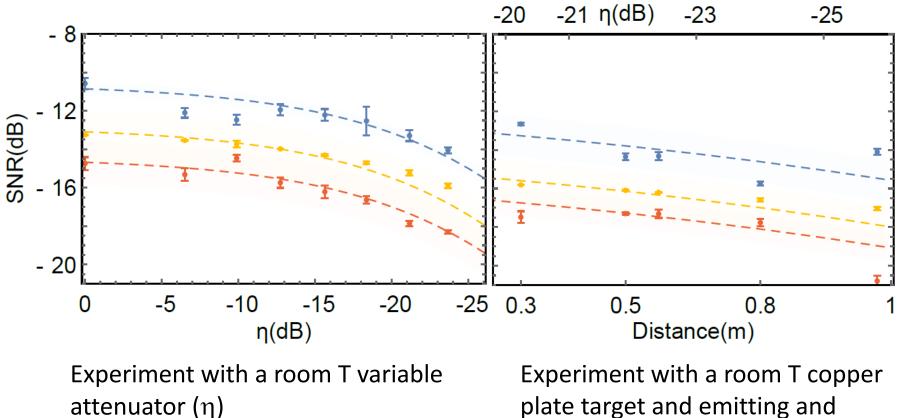
• The signal mode passes through a room T measurement line with a switch used to select between a digitally controllable attenuator, and a free-space link realized with two antennas and a movable reflective object.

• Digital PCR: data postprocessing with the ideal "calibrated" idler a_{I} , obtained rescaling by the measured gain and subtracting the added amplifier noise

Experimental SNR



Target detection



receiving antennas

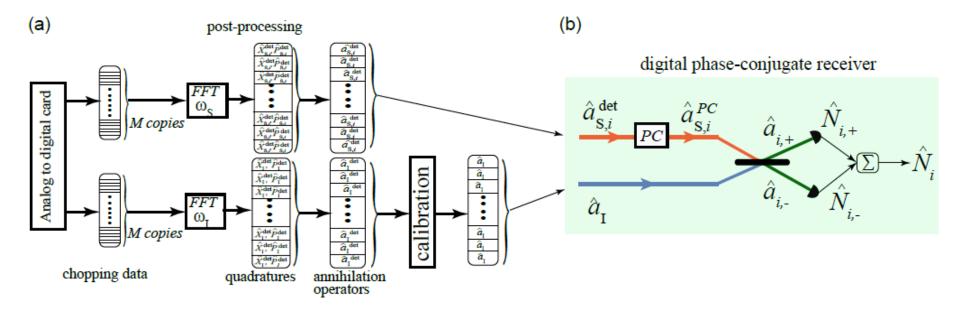
- **1. Blue**: QI = two-mode-squeezed state & digital PCR with calibrated idler
- 2. Orange: coherent state & heterodyne detection
- 3. Red: classical noise radar

Conclusions

- Quantum illumination outperforms any classical target detection strategy, expecially in low-signal/high-noise applications (e.g. radar systems). Up to 6 dB gain in error exponent-rate
- We outperform the classical benchmark of coherent state and homodyne detection by 1 dB at low signal photon number, with entangled signal-idler beams and a digital postprocessed phase-conjugated receiver, at short distance (< 1 meter)
- Potential for short-range radar applications (security, automotive applications, medical imaging)

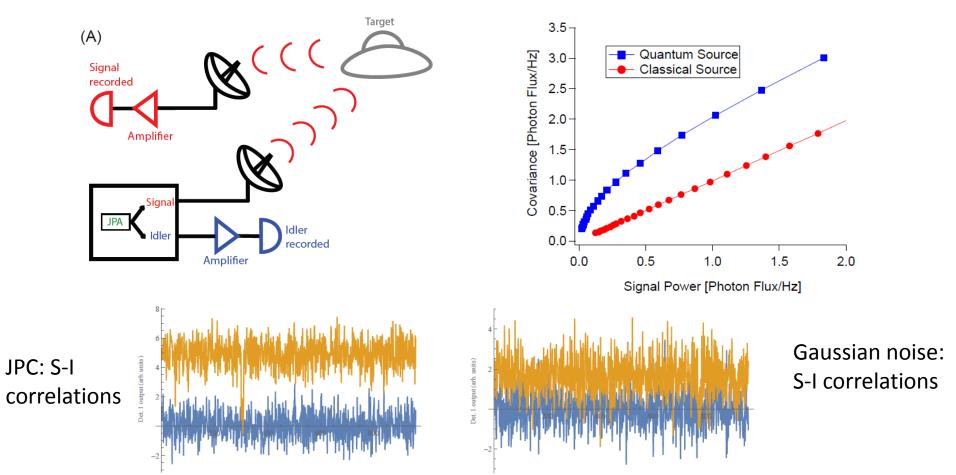
TECHNICAL SLIDES

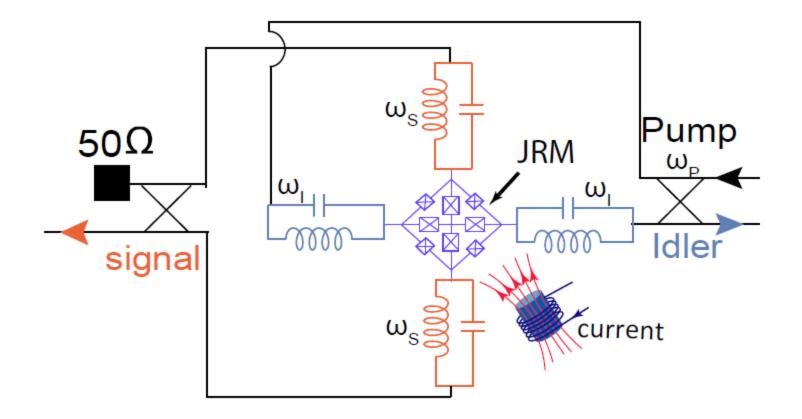
Phase conjugate receiver with digital post-processing



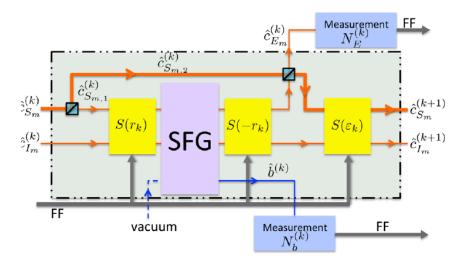
Digital PCR over the reconstructed signal and idler field operators

- Optimal input probe quantum state: two-mode squeezed state from a Josephson parametric converter (JPC).
- Comparison only with a **classical noise radar** (not the optimal classical strategy)
- **Detection is far from optimal**: linear heterodyne measurements and no joint idler-signal measurements





JPC = 4 + 4 Josephson Ring modulators + 3 resonant cavities. Gain 90 dB. 10 and 6.6 GHz

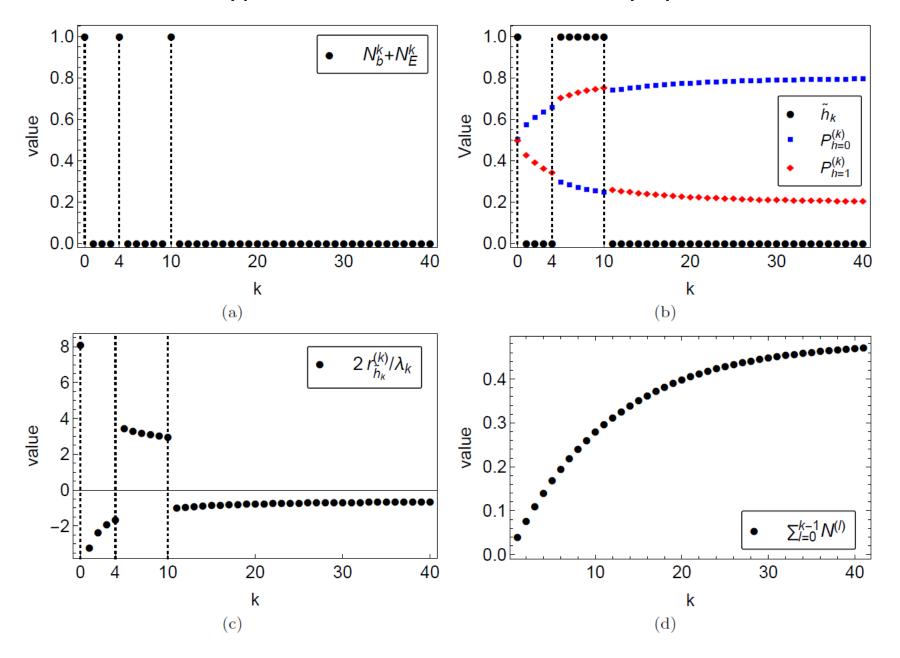


In each run, the squeezing parameters, r_k , ε_k are adjusted according to the results of the k-1 measurements and assuming one of the two options, H_0 or H_1 ,

- 1. The first beam splitter (with very low transmission) makes also the signal with low photon number as needed, (the idler is already low)
- 2. $S(r_k)$ cancels any eventual entanglement between s_m^{k} and I_m^{k}
- If the assumed hypothesis is wrong, b_k≠0, is measured and it is corrected at the next cycle
- Is determined by exploiting the fact that binary decision is equivalent to optimum discrimination between two coherent states (in this case for sum-frequency mode b, even though in a weak thermal environment)
- 5. The second squeezer guarantees that Nb and NE are roughly identical (Nb+NE is the quantity used to check if hypothesis is correct or not)
- 6. The third squeezer has $\varepsilon_k = \sqrt{t} r_k$, where t <<1 input BS transmission, in order to have at the end of each cycle $n^{\text{out}} \sim n^{\text{in}}$

$$n_s^{\text{out}} \simeq n_s^{\text{in}}$$
$$n_i^{\text{out}} \simeq n_i^{\text{in}}$$
$$C_{si}^{\text{out}} \simeq C_{si}^{\text{in}} [1 - \eta (1 + n_s^{\text{in}})].$$

A typical simulation run over many cycles



FURTHER REFINEMENT (Q. Zhuang, Z. Zhang, and J. H. Shapiro, arXiv:1703.02463v1)

- 1. The previous results exploit a **Bayesian approach** in which we assume that the two hypotheses (target present/absent) **have equal probabilities** at the beginning, and then updating the probabilities according to the result of the measurement and Bayes rule. In this context we have minimized the total error probability.
- 2. However, Bayesian analysis is not the preferred approach for target detection, owing to the **difficulty of accurately assigning prior probabilities to target absence and presence** and appropriate costs to false-alarm (Type-I) and miss (Type-II) errors. **Instead, radar theory opts for the Neyman-Pearson performance criterion, in which optimum target detection maximizes the detection probability,** $P_D = Pr(decide present|present), subject to a constraint$ **on the false-alarm probability,** $<math>P_F = Pr(decide present|absent)$. (The detection probability satisfies $P_D = 1 - P_M$, where $P_M = (decide absent|present)$ is the miss probability.)

This is achieved by adapting the previous scheme in the following way:

- 1. We fix a desired (low) value of the false alarm probability P_{F} .
- 2. We then apply the generalized Helstrom optimal binary decision for unequal prior probabilities π_0 and π_1 , which minimizes the error probability (which now means minimizing P_M only), i.e., $P_e^{\min} = \frac{1}{2} \left[1 \frac{1}{2} Tr \left| \left(\hat{\rho}_{RI}^1 \right)^{\otimes M} \frac{\pi_0}{\pi_1} \left(\hat{\rho}_{RI}^0 \right)^{\otimes M} \right| \right]$
- 3. However, we do not know π_0 and π_1 and therefore we start from a ratio π_0/π_1 chosen at will and then we run the protocol in the same way, but with the differently initialized probabilities. It will converge in any case at the end.
- 4. The figure of merit in this Neyman-Pearson scenario is the ROC (receiver operating characteristic), i.e., P_D versus the initially chosen value of P_F.

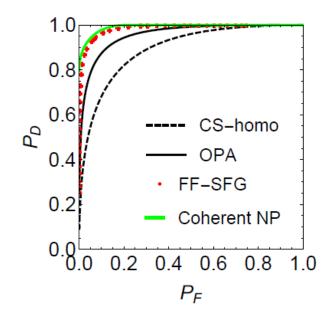


Figure 5. ROCs of QI target detection with FF-SFG reception (red dots), QI target detection with OPA reception (black solid curve), CI target detection with coherent-state (CS) light and homodyne reception (black dashed curve) schemes. Also included is the ROC of coherent-state Neyman-Pearson (Coherent NP) for discriminating between the coherent state $|\sqrt{M\kappa N_S/N_B}\rangle$ and the vacuum state (green solid curve), which is known to be realized by QI target detection with FF-SFG reception when $N_S \ll 1$. All four ROCs assume that $M = 10^{7.5}$, $N_S = 10^{-4}$, $\kappa = 0.01$, and $N_B = 20$.